

## Class 7

- EM wave review
- Calculation of intensity
- Monochromatic Michelson interferometer
- Quasi-monochromatic Michelson
- Autocorrelation theorem
- Fourier Transform interferometer

# Solutions of scalar wave equation

- 2<sup>nd</sup> order PDE:  $\frac{\partial^2}{\partial z^2}\psi(z,t) - \frac{1}{c^2} \frac{\partial^2}{\partial t^2}\psi(z,t) = 0$ 
    - Assume separable solution
    - 2 solutions for  $f(z)$ ,  $g(t)$   $\psi(z,t) = f(z)g(t)$
    - Full solution is a linear combination of both solutions
- $$\psi(z,t) = f(z)g(t) = (A_1 \cos kz + A_2 \sin kz)(B_1 \cos \omega t + B_2 \sin \omega t)$$

- Equivalent representation:

$$\psi(z,t) = A_1 \cos(kz + \omega t + \phi_1) + A_2 \cos(kz - \omega t + \phi_2)$$

forward propagating + backward propagating waves

- Complex (phasor) representation:

$$\psi(z,t) = \text{Re} \left[ a e^{i(kz - \omega t + \phi)} \right] \quad \text{or} \quad \psi(z,t) = \text{Re} \left[ A e^{i(kz - \omega t)} \right]$$

Here  $A$  is complex, includes phase

# Maxwell's Equations to wave eqn

- The induced polarization,  $\mathbf{P}$ , contains the effect of the medium:

$$\begin{aligned}\vec{\nabla} \cdot \mathbf{E} &= 0 & \vec{\nabla} \times \mathbf{E} &= -\frac{\partial \mathbf{B}}{\partial t} \\ \vec{\nabla} \cdot \mathbf{B} &= 0 & \vec{\nabla} \times \mathbf{B} &= \frac{1}{c^2} \frac{\partial \mathbf{E}}{\partial t} + \mu_0 \frac{\partial \mathbf{P}}{\partial t}\end{aligned}$$

Take the curl:

$$\vec{\nabla} \times (\vec{\nabla} \times \mathbf{E}) = -\frac{\partial}{\partial t} \vec{\nabla} \times \mathbf{B} = -\frac{\partial}{\partial t} \left( \frac{1}{c^2} \frac{\partial \mathbf{E}}{\partial t} + \mu_0 \frac{\partial \mathbf{P}}{\partial t} \right)$$

Use the vector ID:

$$\mathbf{A} \times (\mathbf{B} \times \mathbf{C}) = \mathbf{B}(\mathbf{A} \cdot \mathbf{C}) - \mathbf{C}(\mathbf{A} \cdot \mathbf{B})$$

$$\vec{\nabla} \times (\vec{\nabla} \times \mathbf{E}) = \vec{\nabla}(\vec{\nabla} \cdot \mathbf{E}) - (\vec{\nabla} \cdot \vec{\nabla})\mathbf{E} = -\vec{\nabla}^2 \mathbf{E}$$

$$\vec{\nabla}^2 \mathbf{E} - \frac{1}{c^2} \frac{\partial^2 \mathbf{E}}{\partial t^2} = \mu_0 \frac{\partial^2 \mathbf{P}}{\partial t^2}$$

“Inhomogeneous Wave Equation”

# Maxwell's Equations in a Medium

- The induced polarization,  $\mathbf{P}$ , contains the effect of the medium:

$$\vec{\nabla}^2 \mathbf{E} - \frac{1}{c^2} \frac{\partial^2 \mathbf{E}}{\partial t^2} = \mu_0 \frac{\partial^2 \mathbf{P}}{\partial t^2}$$

- Sinusoidal waves of all frequencies are solutions to the wave equation
- The polarization ( $\mathbf{P}$ ) can be thought of as the driving term for the solution to this equation, so the polarization determines which frequencies will occur.
- For linear response,  $\mathbf{P}$  will oscillate at the same frequency as the input.

$$\mathbf{P}(\mathbf{E}) = \epsilon_0 \chi \mathbf{E}$$

- In nonlinear optics, the induced polarization is more complicated:

$$\mathbf{P}(\mathbf{E}) = \epsilon_0 \left( \chi^{(1)} \mathbf{E} + \chi^{(2)} \mathbf{E}^2 + \chi^{(3)} \mathbf{E}^3 + \dots \right)$$

- The extra nonlinear terms can lead to new frequencies.

# Solving the wave equation: linear induced polarization

For low irradiances, the polarization is proportional to the incident field:

$$\mathbf{P}(\mathbf{E}) = \varepsilon_0 \chi \mathbf{E}, \quad \mathbf{D} = \varepsilon_0 \mathbf{E} + \mathbf{P} = \varepsilon_0 (1 + \chi) \mathbf{E} = \varepsilon \mathbf{E} = n^2 \mathbf{E}$$

In this simple (and most common) case, the wave equation becomes:

$$\vec{\nabla}^2 \mathbf{E} - \frac{1}{c^2} \frac{\partial^2 \mathbf{E}}{\partial t^2} = \frac{1}{c^2} \chi \frac{\partial^2 \mathbf{E}}{\partial t^2} \quad \rightarrow \quad \vec{\nabla}^2 \mathbf{E} - \frac{n^2}{c^2} \frac{\partial^2 \mathbf{E}}{\partial t^2} = 0$$

Using:  $\varepsilon_0 \mu_0 = 1 / c^2$

$$\varepsilon_0 (1 + \chi) = \varepsilon = n^2$$

The electric field is a vector function in 3D, so this is actually 3 equations:

$$\vec{\nabla}^2 E_x(\mathbf{r}, t) - \frac{n^2}{c^2} \frac{\partial^2}{\partial t^2} E_x(\mathbf{r}, t) = 0$$

$$\vec{\nabla}^2 E_y(\mathbf{r}, t) - \frac{n^2}{c^2} \frac{\partial^2}{\partial t^2} E_y(\mathbf{r}, t) = 0$$

$$\vec{\nabla}^2 E_z(\mathbf{r}, t) - \frac{n^2}{c^2} \frac{\partial^2}{\partial t^2} E_z(\mathbf{r}, t) = 0$$

# Plane wave solutions for the wave equation

If we assume the solution has no dependence on x or y:

$$\vec{\nabla}^2 \mathbf{E}(z,t) = \frac{\partial^2}{\partial x^2} \mathbf{E}(z,t) + \frac{\partial^2}{\partial y^2} \mathbf{E}(z,t) + \frac{\partial^2}{\partial z^2} \mathbf{E}(z,t) = \frac{\partial^2}{\partial z^2} \mathbf{E}(z,t)$$

$$\rightarrow \frac{\partial^2 \mathbf{E}}{\partial z^2} - \frac{n^2}{c^2} \frac{\partial^2 \mathbf{E}}{\partial t^2} = 0$$

The solutions are oscillating functions, for example

$$\mathbf{E}(z,t) = \hat{\mathbf{x}} E_x \cos(k_z z - \omega t)$$

Where  $\omega = kc$ ,  $k = 2\pi n / \lambda$ ,  $v_{ph} = c / n$

This is a *linearly* polarized wave.

For a plane wave  $\mathbf{E}$  is perpendicular to  $\mathbf{k}$ , so  $\mathbf{E}$  can also point in y-direction

# Complex notation for EM waves

- Write cosine in terms of exponential

$$\mathbf{E}(z,t) = \hat{\mathbf{x}} E_x \cos(kz - \omega t + \phi) = \hat{\mathbf{x}} E_x \frac{1}{2} \left( e^{i(kz - \omega t + \phi)} + e^{-i(kz - \omega t + \phi)} \right)$$

– Note E-field is a *real* quantity.

- It is convenient to work with just one component

– Method 1:  $\mathbf{E}(z,t) = \hat{\mathbf{x}} \operatorname{Re} \left[ A e^{i(kz - \omega t)} \right] \quad A = E_x e^{i\phi}$

– Method 2:  $\mathbf{E}(z,t) = \hat{\mathbf{x}} \left( A e^{i(kz - \omega t)} + c.c. \right) \quad A = \frac{1}{2} E_x e^{i\phi}$

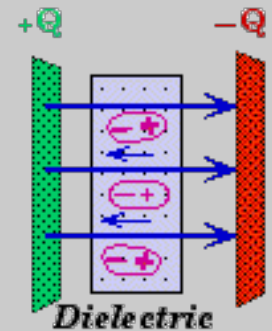
- In *nonlinear* optics, we have to explicitly include conjugate term. Leads to extra factor of  $\frac{1}{2}$ .

# Wave energy and intensity

- Both E and H fields have a corresponding energy density (J/m<sup>3</sup>)
  - For static fields (e.g. in capacitors) the energy density can be calculated through the work done to set up the field

$$\rho = \frac{1}{2}\epsilon E^2 + \frac{1}{2}\mu H^2$$

- Some work is required to polarize the medium
- Energy is contained in both fields, but H field can be calculated from E field

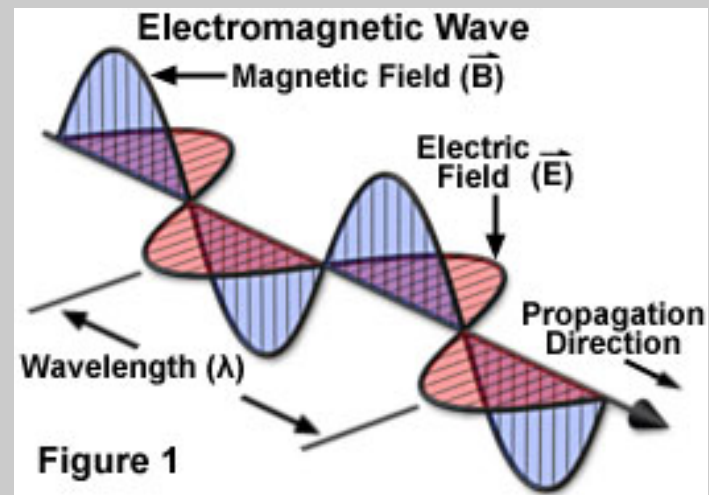




## H field from E field

- H field for a propagating wave is *in phase* with E-field

$$\begin{aligned}\mathbf{H} &= \hat{\mathbf{y}} H_0 \cos(k_z z - \omega t) \\ &= \hat{\mathbf{y}} \frac{k_z}{\omega \mu_0} E_0 \cos(k_z z - \omega t)\end{aligned}$$



- Amplitudes are not independent

$$H_0 = \frac{k_z}{\omega \mu_0} E_0 \quad k_z = n \frac{\omega}{c} \quad c^2 = \frac{1}{\mu_0 \epsilon_0} \rightarrow \frac{1}{\mu_0 c} = \epsilon_0$$

$$H_0 = \frac{n}{c \mu_0} E_0 = n \epsilon_0 c E_0$$

# Energy density in an EM wave

- Back to energy density, non-magnetic

$$\rho = \frac{1}{2} \epsilon E^2 + \frac{1}{2} \mu_0 H^2 \qquad H = n \epsilon_0 c E$$

$$\epsilon = \epsilon_0 n^2$$

$$\rho = \frac{1}{2} \epsilon_0 n^2 E^2 + \frac{1}{2} \mu_0 n^2 \epsilon_0^2 c^2 E^2$$

$$\mu_0 \epsilon_0 c^2 = 1$$

$$\rho = \epsilon_0 n^2 E^2 = \epsilon_0 n^2 E^2 \cos^2(k_z z - \omega t)$$

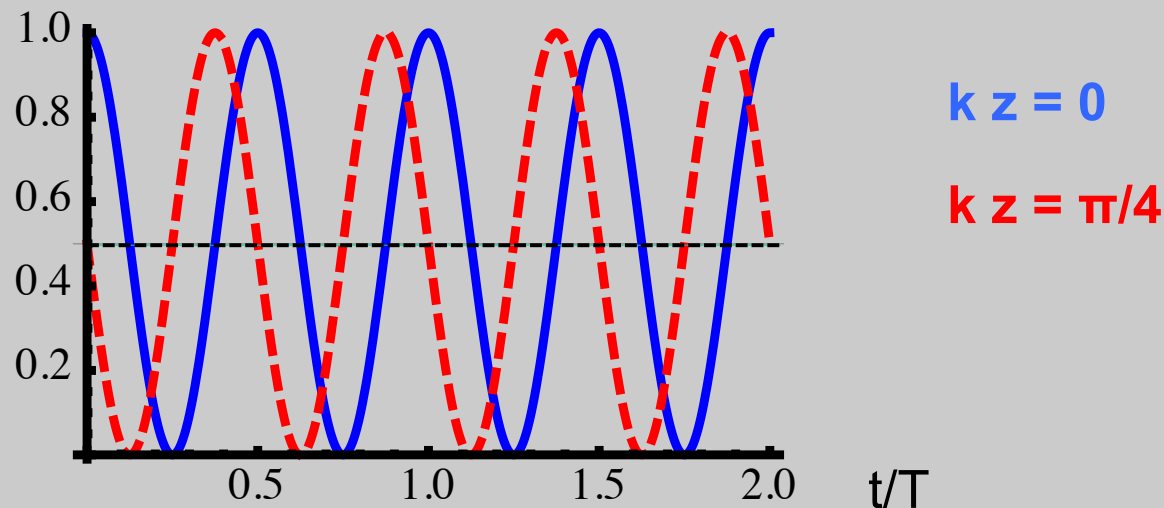
**Equal energy** in both components of wave

# Cycle-averaged energy density

- Optical oscillations are faster than detectors
- Average over one cycle:

$$\langle \rho \rangle = \varepsilon_0 n^2 E_0^2 \frac{1}{T} \int_0^T \cos^2(k_z z - \omega t) dt$$

– Graphically, we can see this should =  $\frac{1}{2}$



– Regardless of position  $z$

$$\langle \rho \rangle = \frac{1}{2} \varepsilon_0 n^2 E_0^2$$

# Intensity and the Poynting vector

- Intensity is an energy flux (J/s/cm<sup>2</sup>)
- In EM the Poynting vector give energy flux

$$\mathbf{S} = \mathbf{E} \times \mathbf{H}$$

– For our plane wave,

$$\mathbf{S} = \mathbf{E} \times \mathbf{H} = E_0 \cos(k_z z - \omega t) n \epsilon_0 c E_0 \cos(k_z z - \omega t) \hat{\mathbf{x}} \times \hat{\mathbf{y}}$$

$$\mathbf{S} = n \epsilon_0 c E_0^2 \cos^2(k_z z - \omega t) \hat{\mathbf{z}}$$

–  $\mathbf{S}$  is along  $\mathbf{k}$

- Time average:  $\mathbf{S} = \frac{1}{2} n \epsilon_0 c E_0^2 \hat{\mathbf{z}}$
- *Intensity* is the magnitude of  $\mathbf{S}$

$$I = \frac{1}{2} n \epsilon_0 c E_0^2 = \frac{c}{n} \rho = V_{phase} \cdot \rho$$

Photon flux:  $F = \frac{I}{h\nu}$

# Calculating intensity with complex wave representation

- Using the convention that we work with the complex form, with the field being the real part

$$\mathbf{E}(z,t) = \hat{\mathbf{x}} \operatorname{Re} \left[ A e^{i(kz - \omega t)} \right] \quad A = E_x e^{i\phi}$$

– Or write

$$\mathbf{E}(z,t) = \mathbf{E}_0 e^{i(kz - \omega t)} \quad \mathbf{E}_0 \text{ complex, vector}$$

– take the real part when we want the *field*

- Time-averaged intensity

$$I = \frac{1}{2} n \epsilon_0 c \mathbf{E}_0 \cdot \mathbf{E}_0^*$$

– Notice this is the sum of intensities for the different polarization components

# Example: Michelson interferometer

- calculate output intensity
  - 50-50 beamsplitter for *power*
  - Transmitted field:

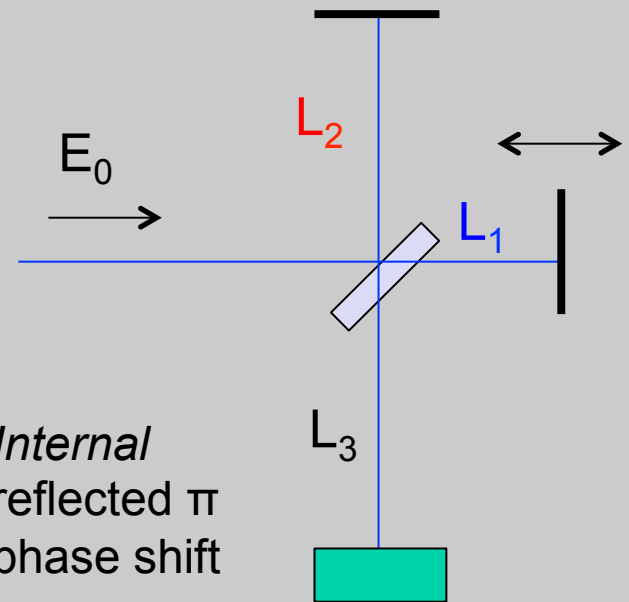
- b/s  $\frac{1}{\sqrt{2}} \hat{\mathbf{x}} E_0 e^{-i\omega t}$
- Return  $\frac{1}{\sqrt{2}} \hat{\mathbf{x}} E_0 e^{i(2kL_1 - \omega t)}$
- Detector  $-\frac{1}{2} \hat{\mathbf{x}} E_0 e^{i[k(2L_1 + L_3) - \omega t]}$

- Reflected field at detector

$$\frac{1}{2} \hat{\mathbf{x}} E_0 e^{i[k(2L_2 + L_3) - \omega t]}$$

- Total field at detector

$$\begin{aligned} \mathbf{E}_{out} &= -\frac{1}{2} \hat{\mathbf{x}} E_0 e^{i[k(2L_1 + L_3) - \omega t]} + \frac{1}{2} \hat{\mathbf{x}} E_0 e^{i[k(2L_2 + L_3) - \omega t]} \\ &= \frac{1}{2} \hat{\mathbf{x}} E_0 e^{i[kL_3 - \omega t]} \left( -e^{ik2L_1} + e^{ik2L_2} \right) \end{aligned}$$



# Michelson: output intensity

- Calculate intensity of output

$$I = \frac{1}{2} n \epsilon_0 c \mathbf{E}_{out} \cdot \mathbf{E}_{out}^* = \frac{1}{2} n \epsilon_0 c \left( |\mathbf{E}_1|^2 + |\mathbf{E}_2|^2 + \mathbf{E}_1 \cdot \mathbf{E}_2^* + \mathbf{E}_2 \cdot \mathbf{E}_1^* \right)$$

$$\mathbf{E}_{out} = \frac{1}{2} \hat{\mathbf{x}} E_0 e^{i[kL_3 - \omega t]} \left( -e^{ik2L_1} + e^{ik2L_2} \right)$$

$$I = \frac{1}{2} n \epsilon_0 c \left( \frac{1}{2} \hat{\mathbf{x}} E_0 e^{i[kL_3 - \omega t]} \left( -e^{ik2L_1} + e^{ik2L_2} \right) \right) \cdot \left( \frac{1}{2} \hat{\mathbf{x}} E_0 e^{i[kL_3 - \omega t]} \left( -e^{ik2L_1} + e^{ik2L_2} \right) \right)^*$$

$$I = \frac{1}{8} n \epsilon_0 c |E_0|^2 \left( -e^{ik2L_1} + e^{ik2L_2} \right) \cdot \left( -e^{-ik2L_1} + e^{-ik2L_2} \right)$$

In terms of input intensity  $I_0 = \frac{1}{2} n \epsilon_0 c |E_0|^2$

$$I_{out} = \frac{1}{4} I_0 \left( 2 - e^{ik2(L_1 - L_2)} - e^{-ik2(L_1 - L_2)} \right)$$

$$= \frac{1}{2} I_0 \left( 1 - \cos \left[ k 2(L_1 - L_2) \right] \right)$$

In terms of *time delay*

$$2k(L_1 - L_2) = \omega \frac{2(L_1 - L_2)}{c} = \omega \tau$$

# Michelson: time-dependent fields

- Now consider the case where the field has time dependence

$$\mathbf{E}_{in}(t) = \hat{\mathbf{x}} E_0(t) e^{-i\omega_0 t} \quad \rightarrow \quad \mathbf{E}_{out}(t) = \frac{1}{2} (\mathbf{E}_{in}(t) - \mathbf{E}_{in}(t - \tau))$$

$$I(t) = \frac{1}{2} n \epsilon_0 c \left( |\mathbf{E}_{in}(t)|^2 + |\mathbf{E}_{in}(t - \tau)|^2 + \mathbf{E}_{in}(t) \cdot \mathbf{E}_{in}(t - \tau)^* + \mathbf{E}_{in}(t - \tau) \cdot \mathbf{E}_{in}(t)^* \right)$$

- This implicitly is a time average over the fast timescale of the carrier

- Now average over a much longer time

$$\langle I(t) \rangle = \int_{-\infty}^{\infty} I(t) dt = 2I_0 + \int_{-\infty}^{\infty} E_0(t) E_0(t - \tau)^* dt + c.c.$$

This part is the field autocorrelation  $E_{AC}(\tau) = \int_{-\infty}^{\infty} E_0(t) E_0^*(t + \tau) dt$   
 $E_{AC}$  is an even function of  $\tau$ , so let  $\tau = -\tau$



# Autocorrelation (Wiener-Khinchin) theorem

$$f_{AC}(\tau) = \int f(t) f^*(t + \tau) dt \quad \text{autocorrelation}$$

- Connect the autocorrelation to the spectrum

$$\begin{aligned} FT_{\tau} \left\{ \int f(t) f^*(t + \tau) dt \right\} &= \iint f(t) f^*(t + \tau) dt e^{i\omega\tau} d\tau \\ &= \int f(t) dt \int f^*(t + \tau) e^{i\omega\tau} d\tau = \int f(t) dt \left[ \int f(t + \tau) e^{-i\omega\tau} d\tau \right]^* \end{aligned}$$

Let  $t' = t + \tau$        $dt' = d\tau$       But flip limits

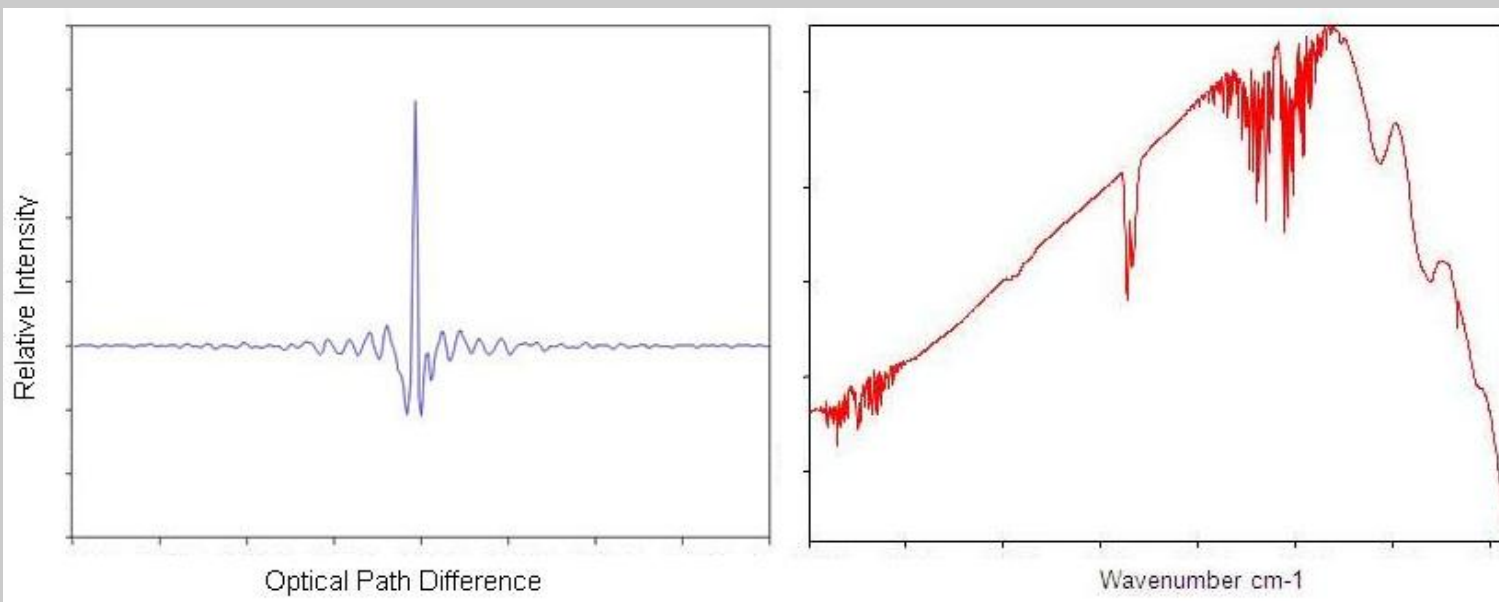
$$\begin{aligned} FT_{\tau} \left\{ f_{AC}(t) \right\} &= \int f(t) dt \left[ \int f(t') e^{-i\omega(t'-t)} dt' \right]^* = \int f(t) dt \left[ F(-\omega) \right]^* e^{-i\omega t} \\ &= F^*(-\omega) \int f(t) e^{-i\omega t} dt = F^*(-\omega) F(-\omega) \end{aligned}$$

If  $f(t)$  is real, then  $F(\omega)$  is even, and

$$FT_{\tau} \left\{ f_{AC}(t) \right\} = |F(\omega)|^2$$

# Fourier transform spectrometer

- Measure interference, subtract DC, FT to get spectrum
  - Single detector, better signal/noise



[http://chemwiki.ucdavis.edu/Physical\\_Chemistry/Spectroscopy/Vibrational\\_Spectroscopy/Infrared\\_Spectroscopy/How\\_an\\_FTIR\\_Spectrometer\\_Operates](http://chemwiki.ucdavis.edu/Physical_Chemistry/Spectroscopy/Vibrational_Spectroscopy/Infrared_Spectroscopy/How_an_FTIR_Spectrometer_Operates)

# Coherence time

- Note that for large time delay, time averaged signal is constant (sum of two intensities)
- Beyond “coherence time” no interference
- Coherence time is inverse of spectral bandwidth

$$T_c \equiv 1 / \Delta\nu$$

